

1000-MHz Quadrature Modulator

Description

U2790B is a 1000-MHz quadrature modulator that uses TELEFUNKEN's advanced UHF process. It features a frequency range from 100 MHz up to 1000 MHz, low current consumption, and single-ended RF and LO ports.

Adjustment free application makes the direct converter suitable for all digital radio systems up to 1000 MHz, e.g., GSM, ADC, JDC.

Features

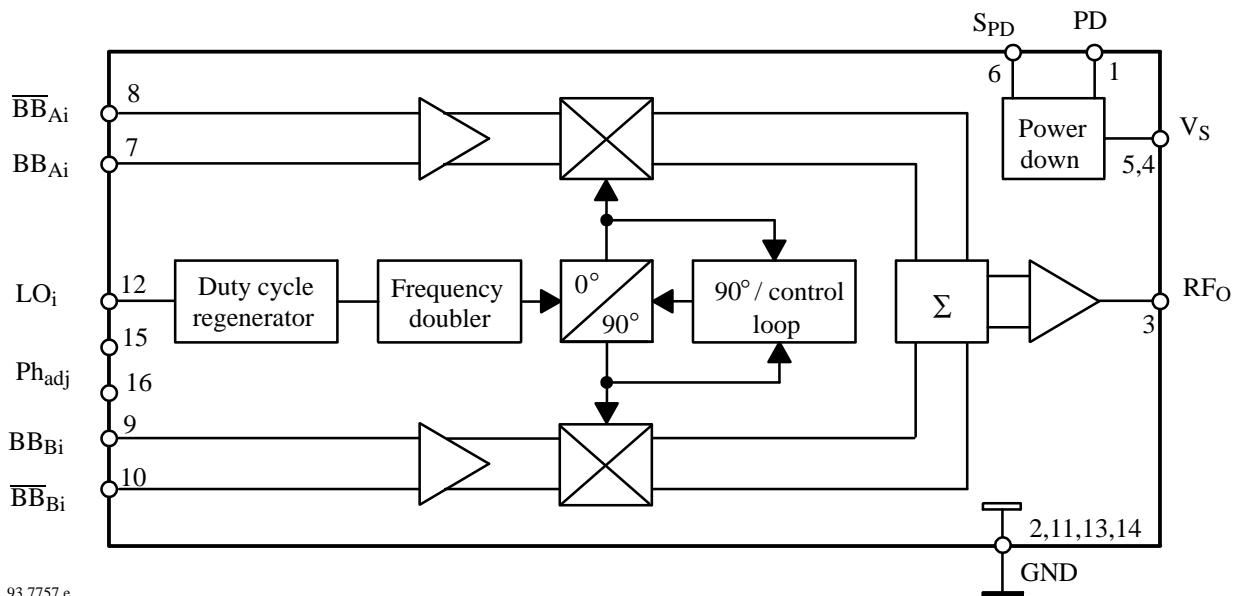
- Supply voltage 5 V (typical)
- Very low power consumption: 150 mW (typical) for -1 dBm output level
- Very good sideband suppression by means of duty cycle regeneration of the LO input signal
- Phase control loop for precise 90° phase shifting
- Power down mode
- Low LO input level: -10 dBm (typical)
- 50-Ω single-ended LO and RF port
- LO- frequency from 100 MHz to 1 GHz
- SO 16 package

Benefits

- No external components required for phase shifting
- Adjustment free, hence saves time
- Only three external components result in cost and board space saving

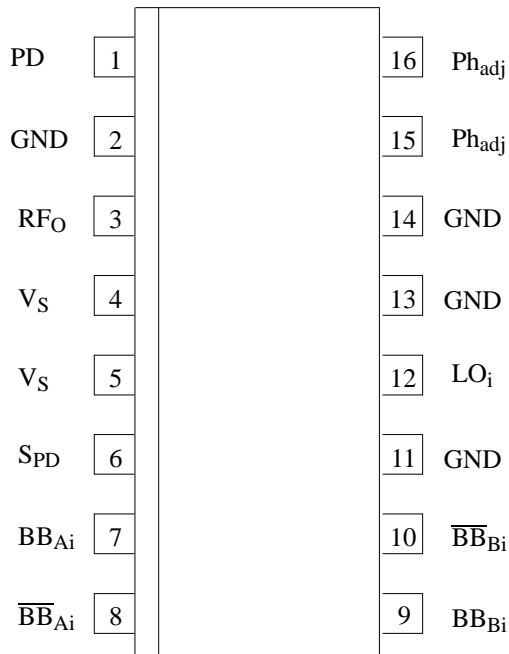
Block Diagram

U2790B-FP



Pin Description

SO 16



94 8023 e

U2790B-FP (SO 16)

Pin	Symbol	Function
1	PD	Power down port
2, 11, 13, 14	GND	Ground
3	RF _o	RF output
4, 5	V _S	Supply voltage
6	S _{PD}	Settling time power down
7	BB _{Ai}	Baseband input A
8	$\overline{\text{BB}}_{\text{Ai}}$	Baseband input A inverse
9	BB _{Bi}	Baseband input B
10	$\overline{\text{BB}}_{\text{Bi}}$	Baseband input B inverse
12	LO _i	LO input
15/16	Ph _{adj}	Phase adjustment (not necessary for regular applications)

Absolute Maximum Ratings

Parameters	Symbol	Value	Unit
Supply voltage Pins 4 and 5	V _S	6	V
Input voltage Pins 7, 8, 9, 10 and 12	V _i	0 to V _S	V
Junction temperature	T _j	125	°C
Storage temperature range	T _{stg}	-40 to +125	°C

Operating Range

Parameters	Symbol	Value	Unit
Supply voltage range Pins 4 and 5	V _S	4.5 to 5.5	V
Ambient temperature range	T _{amb}	-40 to +85	°C

Thermal Resistance

Parameters	Symbol	Value	Unit
Junction ambient SO 16	R _{thJA}	110	K/W

Electrical Characteristics

Test conditions (unless otherwise specified): $V_S = 5\text{ V}$, $T_{\text{amb}} = 25^\circ\text{C}$, referred to test circuit, system impedance $Z_O = 50\ \Omega$, $f_{\text{LO}} = 900\text{ MHz}$, $P_{\text{LO}} = -10\text{ dBm}$, $V_{\text{BBi}} = 1\text{ V}_{\text{pp}}$ diff

Parameters	Test Conditions / Pin	Symbol	Min.	Typ.	Max.	Unit
Supply voltage range	Pins 4 and 5	V_S	4.5		5.5	V
Supply current	Pins 4 and 5	I_S		30		mA
Baseband inputs	Pins 7-8, 9-10					
Input voltage range (differential)		V_{BBi}		1000	1500	mV_{pp}
Input impedance (single ended)		Z_{BBi}		3.2		$\text{k}\Omega$
Input frequency range		f_{BBi}	0		200	MHz
Internal bias voltage		V_{BBb}	2.35	2.5	2.65	V
Temperature coefficient		TC_{BB}		0.1	<1	$\text{mV}/^\circ\text{C}$
LO input	Pin 12					
Frequency range		f_{LOi}	100		1000	MHz
Input level ¹		P_{LOi}	-12	-10	-5	dBm
Input impedance		Z_{iLO}		50		Ω
Voltage standing wave ratio		VSWR_{LO}		1.4	2	-
Duty cycle range		DCR_{LO}	0.4		0.6	-
RF output	Pin 3					
Output level		P_{RFo}	-5	-1		dBm
LO suppression ²	$f_{\text{LO}} = 900\text{ MHz}$ $f_{\text{LO}} = 150\text{ MHz}$	LOR_{Fo}	30 32	35 35		dB
Sideband suppression ^{2,3}	$f_{\text{LO}} = 900\text{ MHz}$ $f_{\text{LO}} = 150\text{ MHz}$	SBS_{RFo}	35 30	40 35		dB
Phase error ⁴		P_e		< 1		deg.
Amplitude error		A_e		< ± 0.25		dB
Noise floor	$V_{\text{BBi}} = 2\text{ V}$, $\overline{V}_{\text{BBi}} = 3\text{ V}$ $V_{\text{BBi}} = \overline{V}_{\text{BBi}} = 2.5\text{ V}$	N_{FL}		-132 -144		dBm/Hz
VSWR		VSWR_{RF}		1.6	2	
3rd order baseband harmonic suppression		S_{BBH}	35	45		dB
RF harmonic suppression		S_{RFH}		35		dB
Power down mode						
Supply current	$V_{\text{PD}} \leq 0.5\text{ V}$ Pins 4, 5 $V_{\text{PD}} = 1\text{ V}$	I_{PD}		10	1	μA
Settling time	$C_{\text{SPD}} = 100\text{ pF}$ $C_{\text{LO}} = 100\text{ pF}$ $C_{\text{RFo}} = 1\text{ nF}$ Pin 6 to 3	t_{SPD}		10		μs
Switching voltage	Pin 1					
Power on		V_{PDon}	4			V
Power down		V_{PDdown}			1	V

Note: 1 The required LO level is a function of the LO frequency.

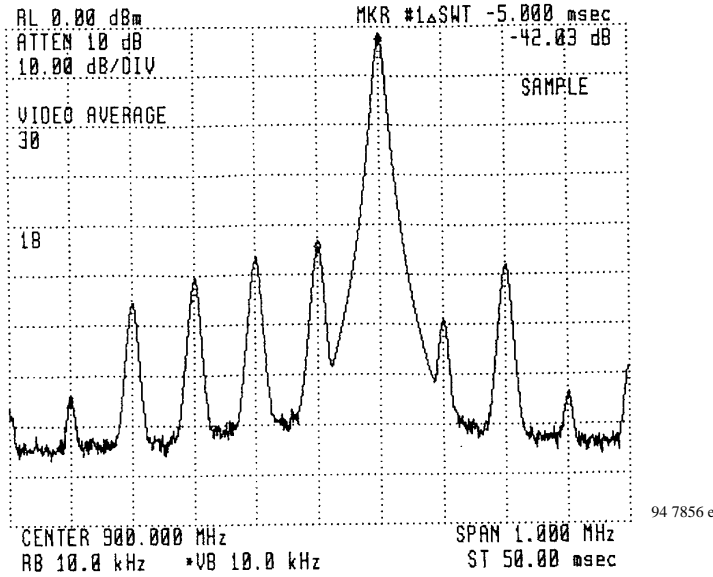
Note: 2 In reference to a RF output level $\leq -1\text{ dBm}$ and I/Q input level of $400\text{ mV}_{\text{pp}}$ diff

Note: 3 Sideband suppression is tested without connection at Pins 15 and 16.

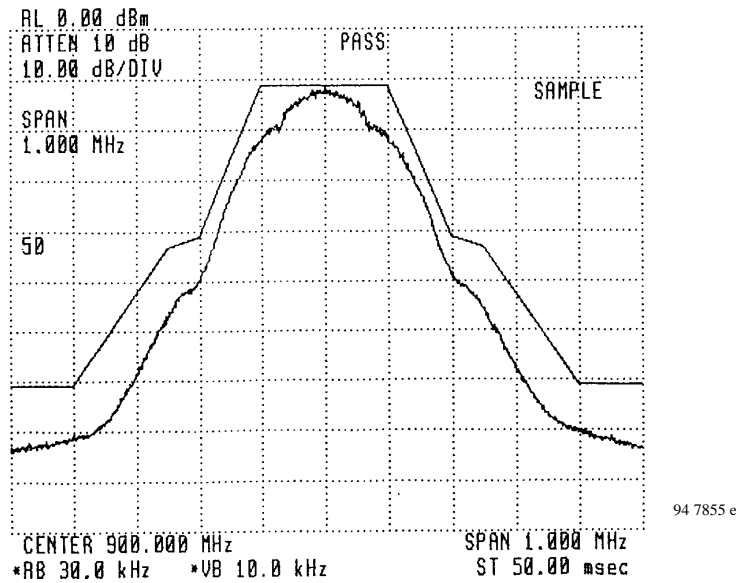
For higher requirements a potentiometer can be connected at these pins.

Note: 4 For $T_{\text{amb}} = -30$ to $+85^\circ\text{C}$ and $V_S = 4.5$ to 5.5 V

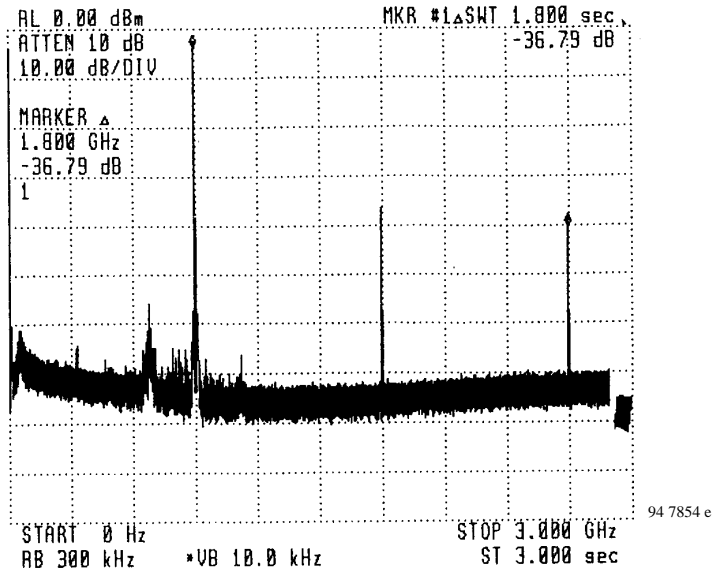
Typical Single Sideband Output Spectrum at $V_S = 4.5\text{ V}$ and $V_S = 5.5\text{ V}$
 $f_{LO} = 900\text{ MHz}$, $P_{LO} = -10\text{ dBm}$, $V_{BBI} = 1\text{ V}_{PP}$ (differential) $T_{amb} = 25^\circ\text{C}$

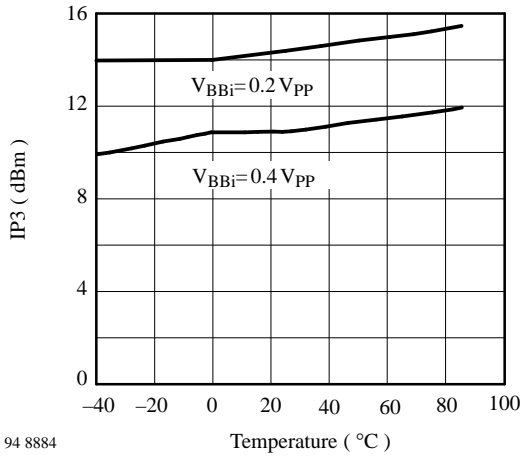


Typical GMSK Output Spectrum

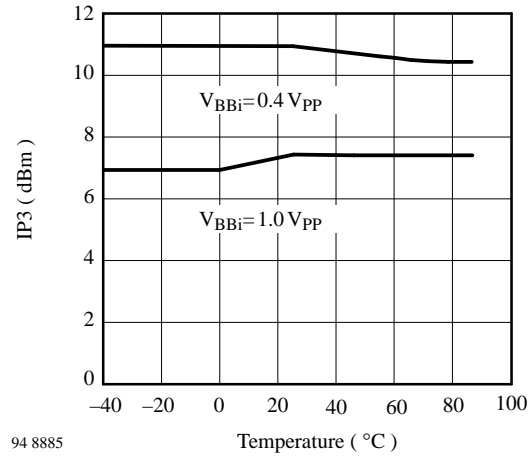


Typical RF-Harmonic Output Spectrum

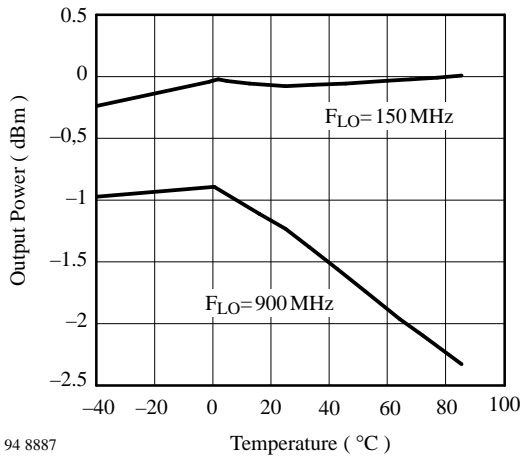




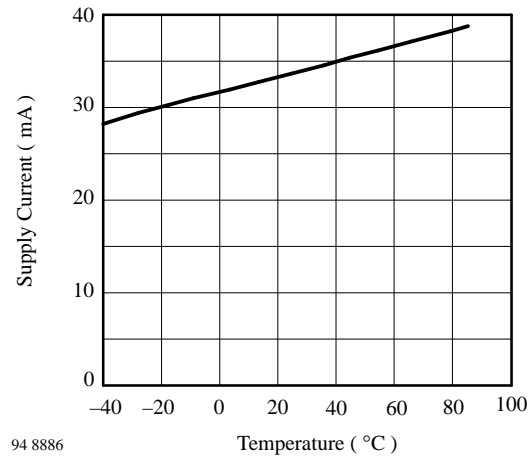
OIP3 vs. T_{amb} , LO = 150 MHz, level - 20 dBm



OIP3 vs. T_{amb} , LO = 900 MHz, level - 10 dBm

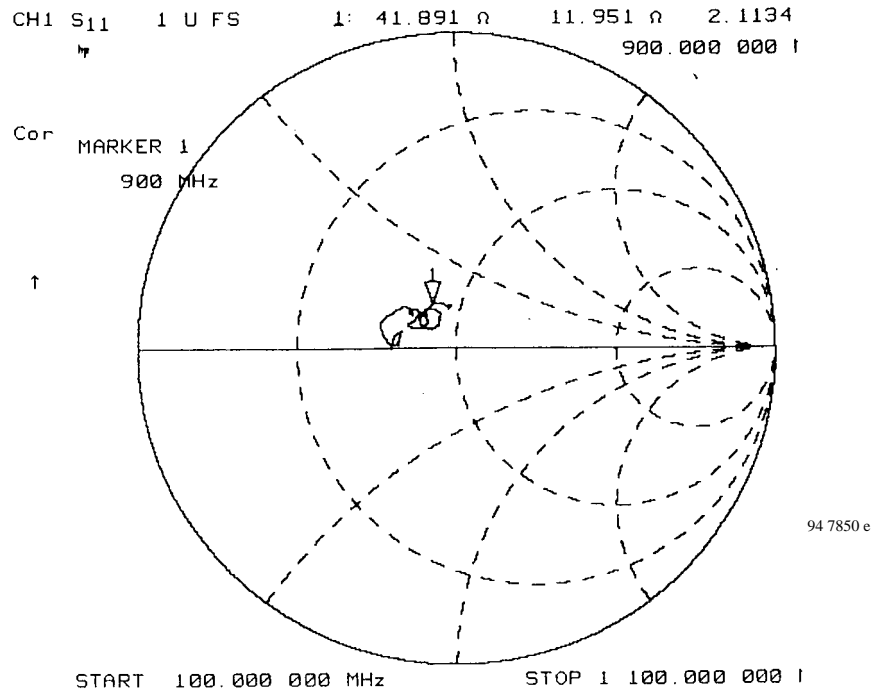


Output power vs. T_{amb}

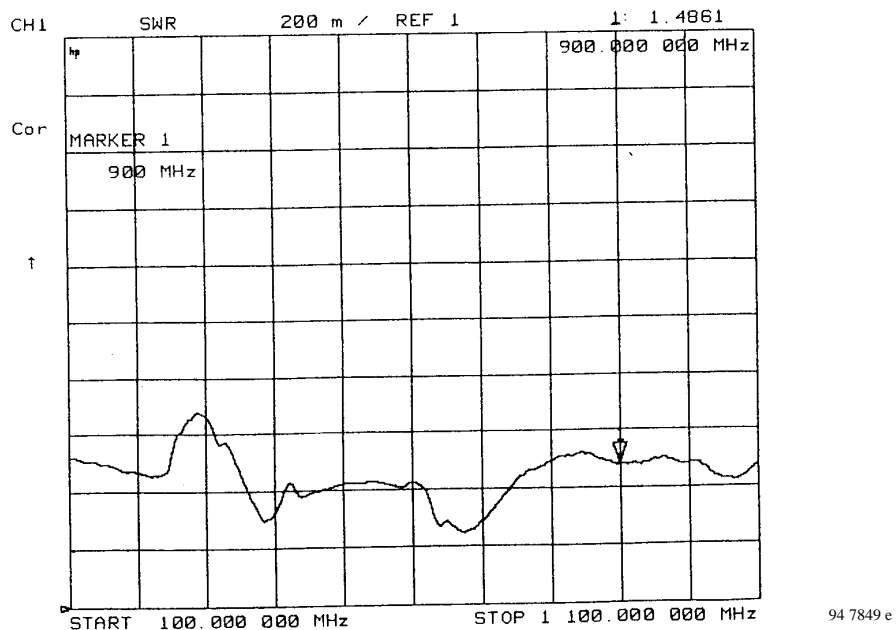


Supply current vs. T_{amb}

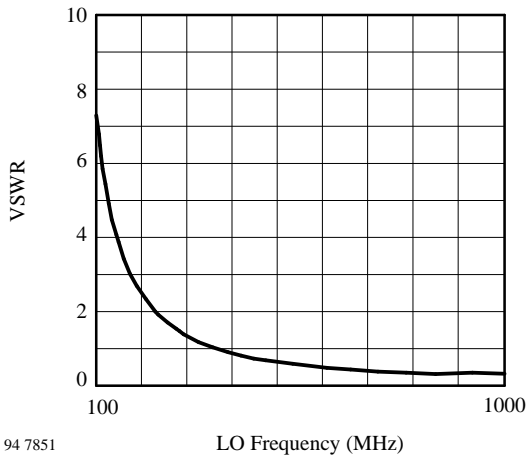
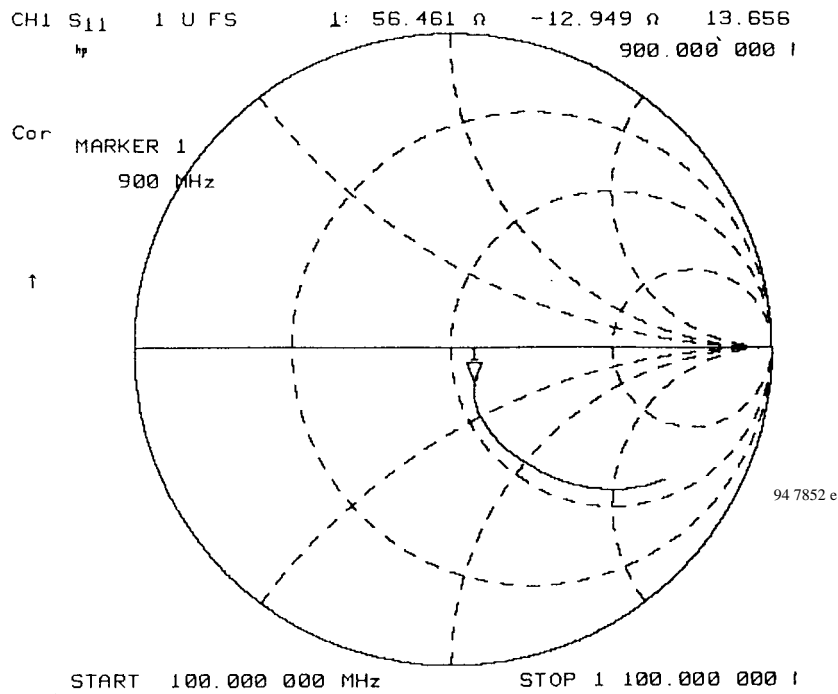
Typical S11 Frequency Response of the RF Output



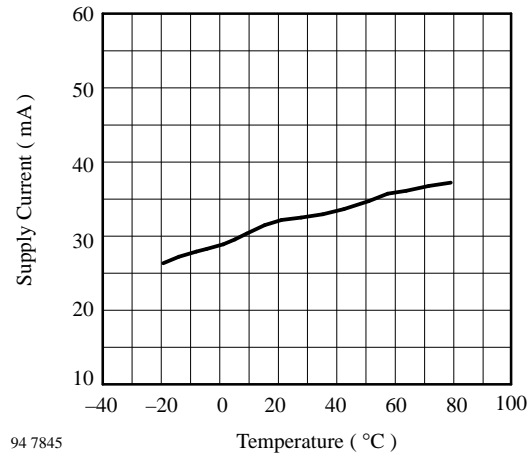
Typical VSWR Frequency Response of the RF Output



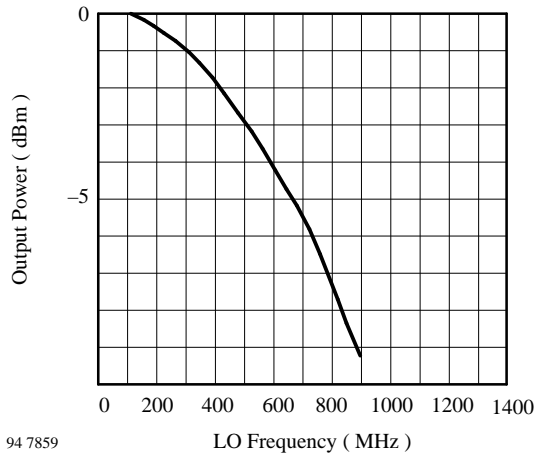
Typical S11 Frequency Response of the LO Input



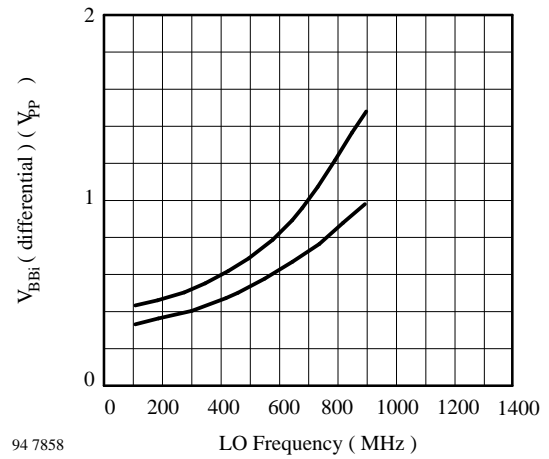
Typical VSWR frequency response of the LO input



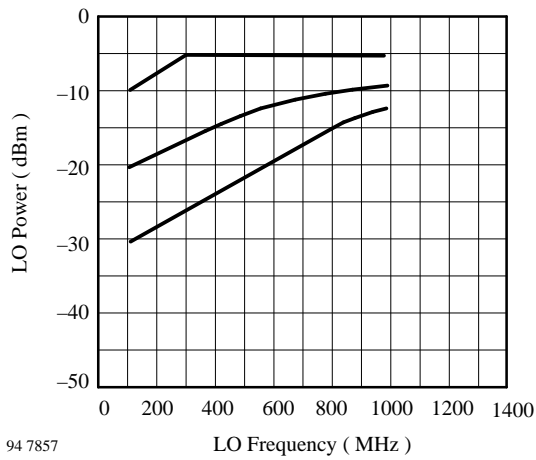
Typical supply current vs. temperature at $V_S = 5\text{ V}$



Typical output power vs. LO-frequency at $T_{amb} = 25^{\circ}C$, $V_{BBI} = 230 \text{ mV}_{PP}$ (differential)

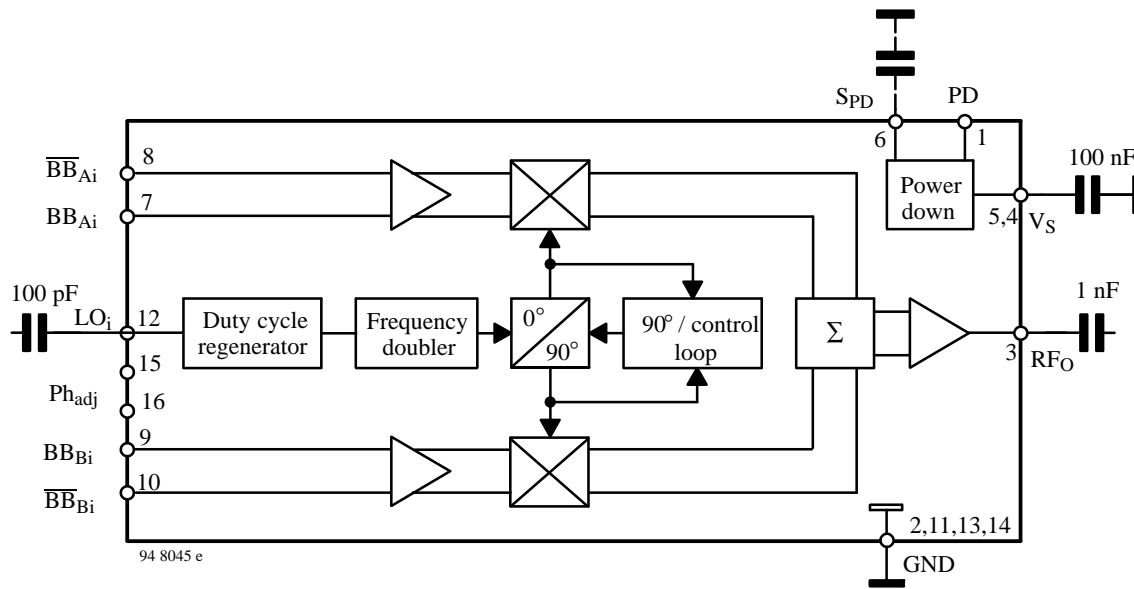


Typical required V_{BBI} input signal (differential) vs. LO frequency for $P_O = 0 \text{ dBm}$ and $P_O = -2 \text{ dBm}$



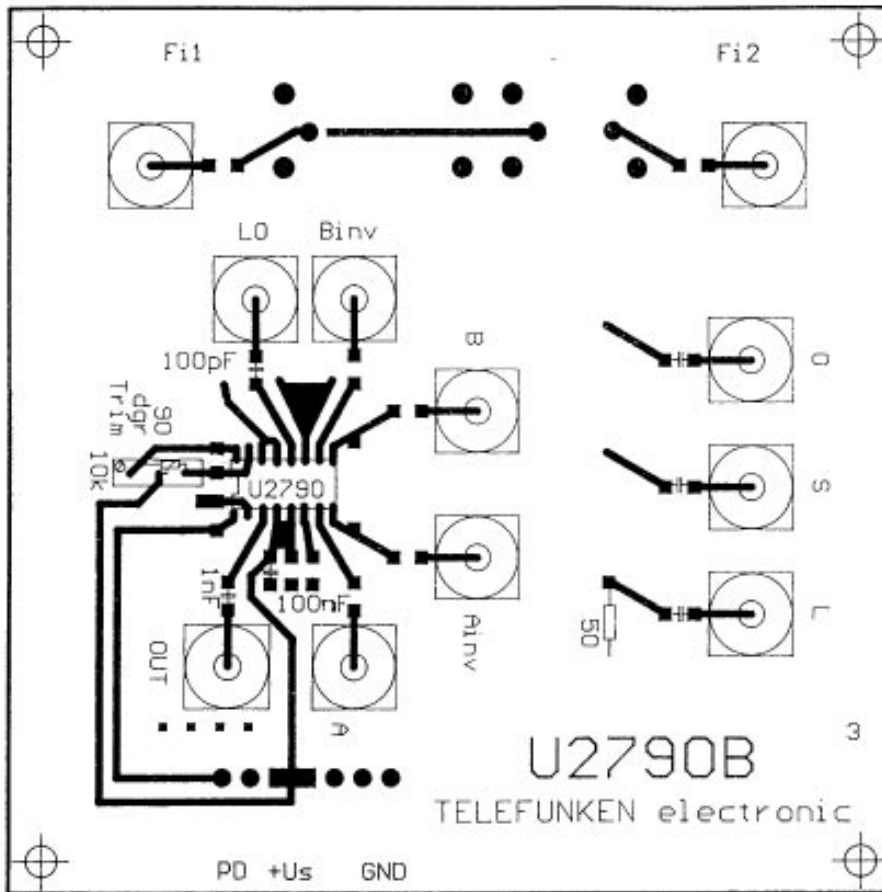
Typical useful LO power range vs. LO frequency at $T_{amb} = 25^{\circ}C$

Application Circuit



PCB Layout

U2790B-FP (SO 16)



Application Notes

1. Noise floor and settling time

In order to reduce noise on the power down control input and improve the wide-off noise floor of the 900-MHz RF output signal, capacitor C_{PD} should be connected from Pin 6 to ground in the shortest possible way.

The settling time has to be considered for the system under design. For GSM applications a value of $C_{PD} = 1 \text{ nF}$ defines a settling time, t_{sPD} , equal or less than $3 \text{ }\mu\text{s}$. This capacitance does not have any influence on the noise floor within the relevant GSM mask. For mobile application the

mask requirements can be achieved very easily without C_{PD} .

A significant improvement of the wide-off noise floor is obtainable with C_{PD} greater than 100 nF . Such values are recommended for applications where the settling time is not critical, such as in base stations. Coupling capacitors for LO_i and RF_O also have a certain impact on the settling time. The values used for the measurements are $C_{LO_i} = 100 \text{ pF}$ and $C_{RF_O} = 1 \text{ nF}$

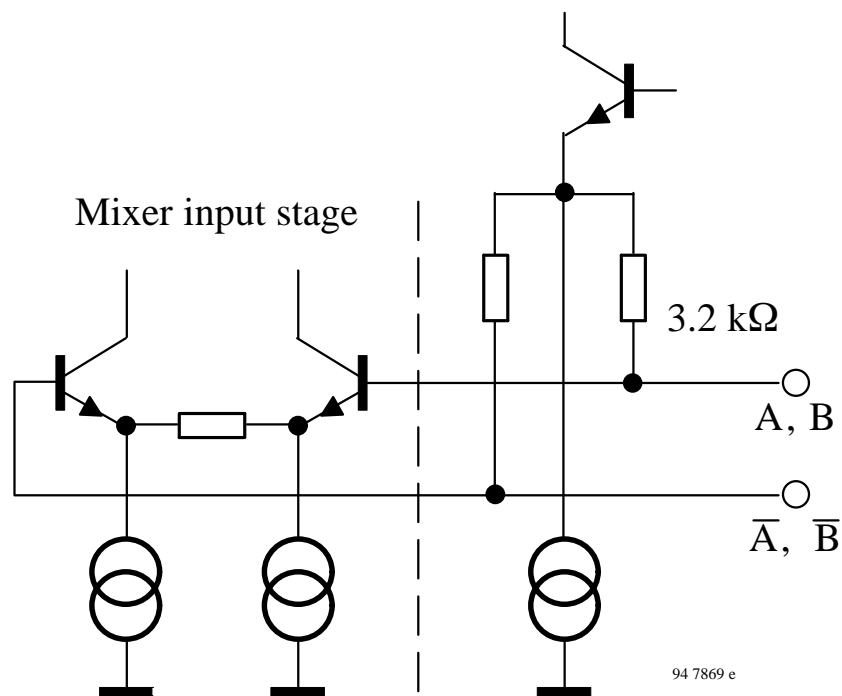


Figure 1 Baseband input circuitry

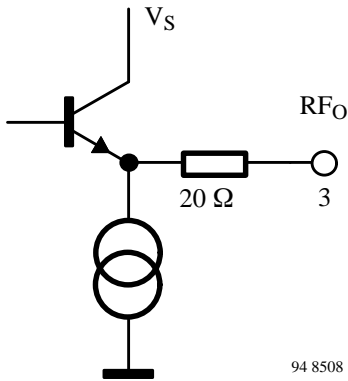
2. Baseband coupling

U2790B-FP (SO 16) has an integrated biasing network which allows ac coupling of the baseband signal at a low count of external components. The bias voltage is $2.5 \text{ V} \pm 0.15 \text{ V}$.

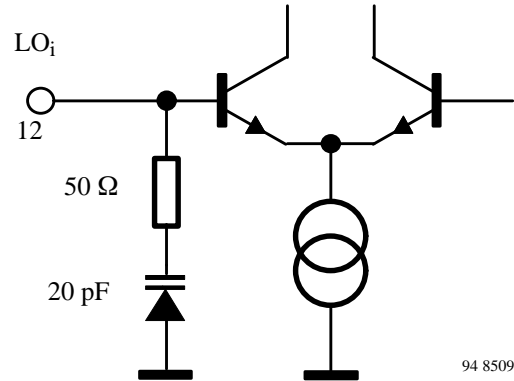
Figure 1 shows the baseband input circuitry with a

resistance of $3.2 \text{ k}\Omega$ for each asymmetric input. The internal dc offset between A and \bar{A} , and B and \bar{B} is typically $< \pm 1 \text{ mV}$ with a maximum of $\pm 3 \text{ mV}$. DC coupling is also possible with an external dc voltage of $2.5 \pm 0.15 \text{ V}$.

3. Circuitries



RF output circuitry



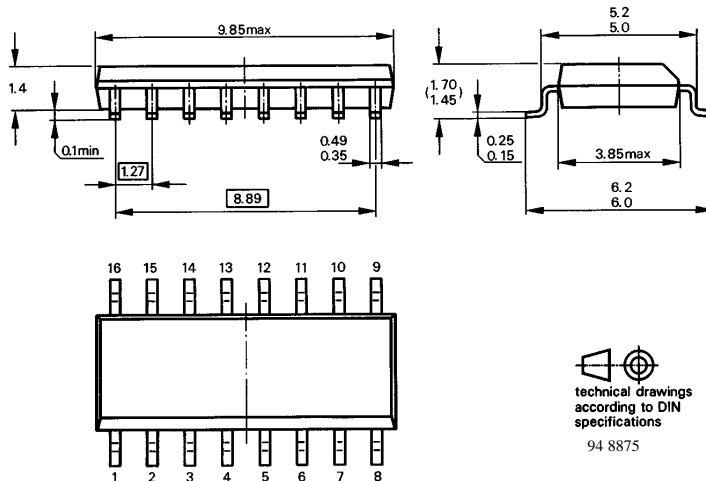
L_O input circuitry

Ordering Information

Package	Part number
SO 16	U2790B-FP

Dimensions in mm

Package: SO 16



We reserve the right to make changes to improve technical design without further notice.

Parameters can vary in different applications. All operating parameters must be validated for each customer application by the customer. Should the buyer use TEMIC products for any unintended or unauthorized application, the buyer shall indemnify TEMIC against all claims, costs, damages, and expenses, arising out of, directly or indirectly, any claim of personal damage, injury or death associated with such unintended or unauthorized use.

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OZONE DEPLETING SUBSTANCES POLICY STATEMENT

It is the policy of **TEMIC TELEFUNKEN microelectronic GmbH** to

1. Meet all present and future national and international statutory requirements and
2. Regularly and continuously improve the performance of our products, processes, distribution and operating systems with respect to their impact on the health and safety of our employees and the public, as well as their impact on the environment.

Of particular concern is the control or elimination of releases into the atmosphere of those substances which are known as ozone depleting substances (ODSs).

The Montreal Protocol (1987) and its London Amendments (1990) will soon severely restrict the use of ODSs and forbid their use within the next ten years. Various national and international initiatives are pressing for an earlier ban on these substances.

TEMIC TELEFUNKEN microelectronic GmbH semiconductor division has been able to use its policy of continuous improvements to eliminate the use of any ODSs listed in the following documents.

1. Annex A, B and list of transitional substances of the Montreal Protocol and the London Amendments respectively
2. Class I and II ozone depleting substances in the Clean Air Act Amendments of 1990 by the Environmental Protection Agency (EPA) in the USA and
3. Council Decision 88/540/EEC and 91/690/EEC Annex A, B and C (transitional substances) respectively.

TEMIC can certify that our semiconductors are not manufactured with and do not contain ozone depleting substances.